Memristors and Nonlinear Programming

Dr. Bharathwaj "Bharath" Muthuswamy

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About me...

 BS (2002), MS (2005), PhD (2009) in EECS from the University of California, Berkeley (advisors: Dr. Leon O. Chua, Dr. Pravin P. Varaiya) Memristors and Nonlinear Programming

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 - For my PhD, my primary contribution was designing, implementing and rigorously proving the existence of chaos in the Muthuswamy-Chua system (circuit): an inductor-capacitor-memristor circuit in series (parallel)

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- Areas of interest:
 - Nonlinear Dynamics (Circuits). Specifically: chaotic circuits and memristors
 - ► Embedded (FPGA) Systems and Education

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 Goal of this talk: Discuss ongoing work from an idea developed at TCNJ - memristor applications to nonlinear programming (specifically, convex optimization) Memristors and Nonlinear Programming

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- ► Organization:
 - ▶ Introduction to the Memristor
 - Fundamental Circuit Variables and Elements
 - Memristor Mathematical Formulation (Gedanken-Experiment)
 - Memristive Models of RRAM (Resistive Random Access Memory)

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 - ► The Memristor Crossbar Array
 - Memristors and Nonlinear Programming
 - Introduction to Nonlinear Programming
 - Introduction to ADMM (Alternative Direction Method of Multipliers)

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► ADMM Using the Memristor Crossbar Array

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- ADMM Using the Memristor Crossbar Array
- Analog emulation of Memristor RRAM

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- ADMM Using the Memristor Crossbar Array
- Analog emulation of Memristor RRAM
- ► Conclusion and Q/A

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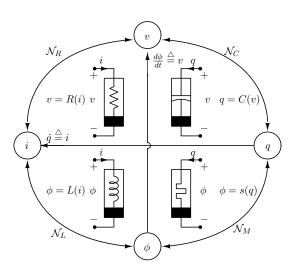
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Fundamental Circuit Variables and Elements



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A memristor (menductor) defines a relationship between ϕ and q (q and ϕ):

$$\phi \stackrel{\triangle}{=} s(q) \tag{1}$$

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$$\phi \stackrel{\triangle}{=} s(q) \tag{1}$$

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► In terms of current and voltage:

$$v(t) = M(q(t))i(t) \tag{2}$$

Here, $M(q(t)) = M(\int_{-\infty}^t i(\tau) d\tau).$ M is the memristance function

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- ► Eq. (2) defines an ideal memristor (Josephson junctions)
- Generalization to a nonideal memristor:

$$v(t) = M(\mathbf{x}(t))i(t)$$

$$\dot{\mathbf{x}}(t) = f(\mathbf{x}, i)$$
(3)

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Discharge tubes (internal state x = number of conduction electrons n):

$$v = M(n)i (4)$$

$$\frac{dn}{dt} = -\beta n + \alpha M(n)i^2 \tag{5}$$

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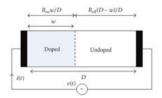
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► RRAM (ReRAM) memristive models were popularized by HP (2008)



$$M(w(t)) = R_{on} \frac{w(t)}{D} + R_{off} \left(\frac{D - w(t)}{D}\right)$$
 (6)

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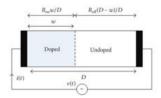
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 Concept of forming and destroying conductive filaments through an insulating material Memristors and Nonlinear Programming

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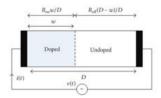
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- Concept of forming and destroying conductive filaments through an insulating material
- For HP RRAM, oxygen vacancies act as conductive filaments

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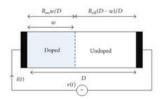
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$$M(w(t)) = R_{on} \frac{w(t)}{D} + R_{off} \left(\frac{D - w(t)}{D}\right)$$
 (6)

4 D > 4 P > 4 B > 4 B > B

- Concept of forming and destroying conductive filaments through an insulating material
- For HP RRAM, oxygen vacancies act as conductive filaments
- Usually, $R_{off} >> R_{on}$

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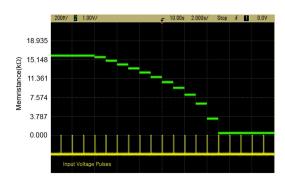
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► Changing memristance for input current pulses



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Potential advantages (as compared to traditional three-terminal transistor based memory): Memristors and Nonlinear Programming

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- Potential advantages (as compared to traditional three-terminal transistor based memory):
 - ► Since a memristor is a two-terminal device, smaller footprint ⇒ higher memory densities

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- Potential advantages (as compared to traditional three-terminal transistor based memory):
 - Since a memristor is a two-terminal device, smaller footprint ⇒ higher memory densities
 - Non-volatile (resistance is retained when v(t)=0) analog (continuous change in resistance values) memory

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- ► Potential advantages (as compared to traditional three-terminal transistor based memory):
 - Since a memristor is a two-terminal device, smaller footprint ⇒ higher memory densities
 - Non-volatile (resistance is retained when v(t)=0) analog (continuous change in resistance values) memory
- However, a variety of fabrication challenges associated with RRAM (discussed in analog emulation of memristors)

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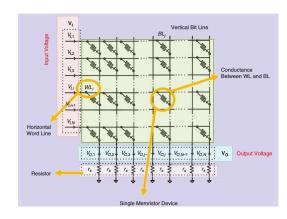
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 \blacktriangleright Output voltage(s) $\mathbf{V_O}$ as a function of input voltage(s) $\mathbf{V_I}$:

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Output voltage(s) V_O as a function of input voltage(s) V_I:

$$\begin{bmatrix} V_{O,1} \\ \vdots \\ V_{O,M} \end{bmatrix} = \begin{bmatrix} \frac{g_{1,1}}{\sum\limits_{S=1}^{N} g_{1,K}} & \frac{g_{1,2}}{\sum\limits_{S=1}^{N} g_{1,K}} & \cdots & \frac{g_{1,N}}{\sum\limits_{S=1}^{N} g_{1,K}} \\ \vdots & \vdots & \ddots & \vdots \\ \frac{g_{M,1}}{g_{s} + \sum\limits_{K=1}^{N} g_{1,K}} & \frac{g_{M,2}}{\sum\limits_{S=1}^{N} g_{1,K}} & \cdots & \frac{g_{M,N}}{\sum\limits_{S=1}^{N} g_{1,K}} \end{bmatrix} \begin{bmatrix} V_{I,1} \\ \vdots \\ V_{I,N} \end{bmatrix}$$

$$V_{O} = CV_{I} \tag{8}$$

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$$V_{O} = CV_{I} \tag{8}$$

Notice that a memristor crossbar array has O(1) time complexity with respect to analog matrix multiplicaton

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• Output voltage(s) V_O as a function of input voltage(s) V_I :

$$\begin{bmatrix} V_{O,1} \\ \vdots \\ V_{O,M} \end{bmatrix} = \begin{bmatrix} \frac{g_{1,1}}{s+\sum_{k=1}^{N} g_{1,K}} & \frac{g_{1,2}}{s+\sum_{k=1}^{N} g_{1,K}} & \cdots & \frac{g_{1,N}}{s+\sum_{k=1}^{N} g_{1,K}} \\ \vdots & \vdots & \ddots & \vdots \\ \frac{g_{M,1}}{g_{s}+\sum_{k=1}^{N} g_{1,K}} & \frac{g_{M,2}}{s+\sum_{k=1}^{N} g_{1,K}} & \cdots & \frac{g_{M,N}}{s+\sum_{k=1}^{N} g_{1,K}} \end{bmatrix} \begin{bmatrix} V_{I,1} \\ \vdots \\ V_{I,N} \end{bmatrix}$$

$$V_{O} = CV_{I} \tag{8}$$

- ► Notice that a memristor crossbar array has O(1) time complexity with respect to analog matrix multiplicaton
- ► Primary challenges in memristor crossbar array technology: scalability (addressed by using arbiters), process-variablity (hot topic of research), negative coefficients...

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► Negative coefficients...

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Negative coefficients...

$$\mathbf{V_0} = \mathbf{CV_I}$$

$$= (\mathbf{C_+} - (-\mathbf{C_-})) \mathbf{V_I}$$

$$= (\mathbf{C_+} - (\mathbf{BD})) \mathbf{V_I}$$
(9)

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► Negative coefficients...

$$\begin{aligned} \mathbf{V_0} &= \mathbf{C}\mathbf{V_I} \\ &= \left(\mathbf{C_+} - \left(-\mathbf{C_-}\right)\right)\mathbf{V_I} \\ &= \left(\mathbf{C_+} - \left(\mathbf{B}\mathbf{D}\right)\right)\mathbf{V_I} \end{aligned} \tag{9}$$

- ► Mathematical "trick" of auxiliary variables in defining:
 - \tilde{N} : number of nonzero columns of $-\mathbf{C}_{-}$
 - $ightharpoonup \mathbf{B} \in \mathbb{R}^{N imes ilde{N}}$: nonzero columns of $-\mathbf{C}_-$
 - $\mathbf{D} \in \mathbb{R}^{\tilde{N} \times N}$: indices of the nonzero columns of $-\mathbf{C}_{-}$

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 - \tilde{N} : number of nonzero columns of $-\mathbf{C}_{-}$
 - $ightharpoonup \mathbf{B} \in \mathbb{R}^{N imes ilde{N}}$: nonzero columns of $-\mathbf{C}_-$
 - $oldsymbol{\mathrm{D}} \in \mathbb{R}^{ ilde{N} imes N}$: indices of the nonzero columns of $-\mathbf{C}_-$
- ▶ Usually $\tilde{N} << N$, size of resulting memristor crossbar array is $(N + \tilde{N}) \times (N + \tilde{N})$

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Example...

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► Example...

$$\mathbf{C} = \begin{bmatrix} 1 & -3 & 0 \\ 0 & -0.75 & 4 \\ -2 & -5 & 6 \end{bmatrix}$$

$$= \begin{bmatrix} 1 & 0 & 0 \\ 0 & 0 & 4 \\ 0 & 0 & 6 \end{bmatrix} - \begin{bmatrix} 0 & 3 & 0 \\ 0 & 0.75 & 0 \\ 2 & 5 & 0 \end{bmatrix}$$

$$= \begin{bmatrix} 1 & 0 & 0 \\ 0 & 0 & 4 \\ 0 & 0 & 6 \end{bmatrix} - \begin{bmatrix} 0 & 3 \\ 0 & 0.75 \\ 2 & 5 \end{bmatrix} \begin{bmatrix} 1 & 0 & 0 \\ 0 & 1 & 0 \end{bmatrix}$$

$$= (\mathbf{C}_{+} - (\mathbf{BD})) \tag{10}$$

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Memristor Crossbar Array Schematic

► Example...

$$\mathbf{C} = \begin{bmatrix} 1 & -3 & 0 \\ 0 & -0.75 & 0 \\ -2 & -5 & 6 \end{bmatrix}$$

$$= \begin{bmatrix} 1 & 0 & 0 \\ 0 & 0 & 4 \\ 0 & 0 & 6 \end{bmatrix} - \begin{bmatrix} 0 & 3 & 0 \\ 0 & 0.75 & 0 \\ 2 & 5 & 0 \end{bmatrix}$$

$$= \begin{bmatrix} 1 & 0 & 0 \\ 0 & 0 & 4 \\ 0 & 0 & 6 \end{bmatrix} - \begin{bmatrix} 0 & 3 \\ 0 & 0.75 \\ 2 & 5 \end{bmatrix} \begin{bmatrix} 1 & 0 & 0 \\ 0 & 1 & 0 \end{bmatrix}$$

$$= (\mathbf{C}_{+} - (\mathbf{B}\mathbf{D})) \tag{10}$$

Now, a variety of nonlinear programming problems can be mapped onto the memristor crossbar array...

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 Nonlinear programming is essentially the concept of finding an optimal solution to a problem, given a set of constraints Memristors and Nonlinear Programming

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- Nonlinear programming is essentially the concept of finding an optimal solution to a problem, given a set of constraints
- ► Example: Consider the problem of sizing devices on an IC (Integrated Circuit) task of choosing the width and length of each device in the IC:

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Analog Emulation of



- Nonlinear programming is essentially the concept of finding an optimal solution to a problem, given a set of constraints
- Example: Consider the problem of sizing devices on an IC (Integrated Circuit) - task of choosing the width and length of each device in the IC:
 - Variables: width, length of each device

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 Constraint functions: timing threshold, device area, total area etc. Memristors and Nonlinear Programming

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Convex Optimization

▶ Mathematically (f_0 : objective function, \mathbf{f}_i : constraint function(s) $f_0, \mathbf{f}_i : \mathbb{R}^n \to \mathbb{R}$):

minimize:
$$f_0(\mathbf{x})$$
 (11)

subject to :
$$\mathbf{f}_i(\mathbf{x}) < \mathbf{b}_i \quad i = 1, \dots, m$$
 (12)

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 $lackbox{ We have a convex optimization problem iff } f_0, \mathbf{f}_i$ are convex:

$$f_k(\alpha \mathbf{x} + \beta \mathbf{y}) \le \alpha f_k(\mathbf{x}) + \beta f_k(\mathbf{y})$$
 (13)

with
$$\alpha + \beta > 1, \alpha > 0, \beta > 0$$

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Introduction to ADMM

 ADMM : Alternative Direction Method of Multipliers decompose optimization variable into two simpler groups Memristors and Nonlinear Programming

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Introduction to ADMM

- ADMM: Alternative Direction Method of Multipliers decompose optimization variable into two simpler groups
- ▶ One of the main steps in ADMM involves solving the optimization problem:

$$\text{minimize}: \frac{\rho}{2} \|\mathbf{x} - \alpha\|_2^2 \tag{14}$$

subject to :
$$Gx = h$$
 (15)

where: $\mathbf{x} \in \mathbb{R}^n$ is the optimization variable, α is a "coupling" parameter, $\mathbf{G} \in \mathbb{R}^{l \times n}$ and $\mathbf{h} \in \mathbb{R}^l$ are parameters.

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Using the Karush-Kuhn-Tucker (KKT) necessary and sufficient conditions, we get the following equivalent equation to solve:

$$\rho(\mathbf{x} - \alpha) + \mathbf{G}^T \lambda = \mathbf{0} \tag{16}$$

$$Gx = h$$
 (17)

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where $\lambda \in \mathbb{R}^l$ is the Lagrange multiplier. Eqs. (16) and (17) can be rewritten as:

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where $\lambda \in \mathbb{R}^l$ is the Lagrange multiplier. Eqs. (16) and (17) can be rewritten as:

$$\begin{bmatrix} \rho \mathbf{I} & \mathbf{G}^T \\ \mathbf{G} & \mathbf{0} \end{bmatrix} \begin{bmatrix} \mathbf{x} \\ \lambda \end{bmatrix} = \begin{bmatrix} \rho \alpha \\ \mathbf{h} \end{bmatrix}$$
 (18)

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 (18)

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The system in Eq. (18) can be efficiently mapped to memristor crossbars by configuring their memristance values according to $\mathbf{C} = \begin{bmatrix} \rho \mathbf{I} & \mathbf{G}^T \\ \mathbf{G} & \mathbf{0} \end{bmatrix}$

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Sample application: Model Predictive Control (MPC) reformulated as a Mixed Linear Complementarity Problem (MLCP) For Autonomous Vehicles! Memristors and Nonlinear Programming

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- Sample application: Model Predictive Control (MPC) reformulated as a Mixed Linear Complementarity Problem (MLCP) For Autonomous Vehicles!
- Proposal for a TCNJ ECE Senior Design Project in AY 2018 - 2019...

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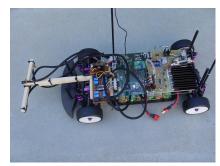
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► NATCAR, Berkeley EECS 192

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- Some major issues with current RRAM technology:
 - Lack of proper models that predict physical device behavior

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- Some major issues with current RRAM technology:
 - Lack of proper models that predict physical device behavior
 - ► Memristive (RRAM) devices fall well short of CMOS in terms of endurance

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- Some major issues with current RRAM technology:
 - Lack of proper models that predict physical device behavior
 - Memristive (RRAM) devices fall well short of CMOS in terms of endurance
- Since memristive RRAM models and devices are still a topic of research...



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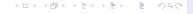
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Conclusio

- Some major issues with current RRAM technology:
 - Lack of proper models that predict physical device behavior
 - Memristive (RRAM) devices fall well short of CMOS in terms of endurance
- Since memristive RRAM models and devices are still a topic of research...
- …there is a need for robust (analog) emulators





CMOS-based Analog Emulation of Memristor RRAM: Block Diagram

Analog emulation of memristors has been done using opamps etc. Memristors and Nonlinear Programming

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CMOS-based Analog Emulation of Memristor RRAM: Block Diagram

- Analog emulation of memristors has been done using opamps etc.
- But we would like solutions that are easy to fabricate in CMOS

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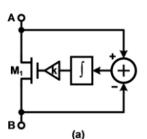
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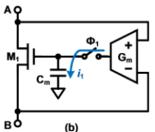




CMOS-based Analog Emulation of Memristor RRAM: Block Diagram

- Analog emulation of memristors has been done using opamps etc.
- ▶ But we would like solutions that are easy to fabricate in **CMOS**





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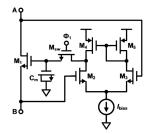
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CMOS-based Analog Emulation of Memristor RRAM: Implementation



$$\begin{split} I &\approx K_{n} \frac{W}{L} \left(V_{G} - V_{C_{M}} - V_{T_{N}} \right) \cdot V_{AB} \\ &= K_{n} \frac{W}{L} \left(x - V_{C_{M}} - V_{T_{N}} \right) \cdot V_{AB} \end{split} \tag{19}$$

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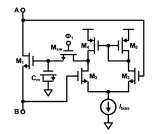
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CMOS-based Analog Emulation of Memristor RRAM: Implementation



$$I \approx K_n \frac{W}{L} \left(V_G - V_{C_M} - V_{T_N} \right) \cdot V_{AB}$$
$$= K_n \frac{W}{L} \left(x - V_{C_M} - V_{T_N} \right) \cdot V_{AB} \tag{19}$$

$$\dot{x} = \frac{G_M(V_{AB})}{C_M}$$

$$= f(V_{AB})$$
(20)

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Conclusion, Q/A and References

- Ongoing research:
 - Implement convex optimization problems, say MLCP, via ADMM on memristor crossbar array (work jointly done with TCNJ faculty (Dr. Adegbege) and students (Paul B., Jake B., Matt K.)):
 - Simulation framework for memristors (crossbar array):
 QUCS (Quite Universal Circuit Simulator), MAPP
 (Berkeley Model and Algorithm Prototyping Platform)
 - Synthesize memristor crossbar array using Analog RRAM memristor emulator
 - Senior design project for AY 2018-2019
 - Memristive models for atrial fibrillation (with Dr. S. T. Mathew)
 - Noise induced chaos in the Muthuswamy-Chua system (with Dr. K. Ganesan and Dr. S. Banerjee)
- ▶ Questions?

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