

Chaotic Dynamics of the Muthuswamy-Chua¹ System

IIT Chennai
August 23rd 2013

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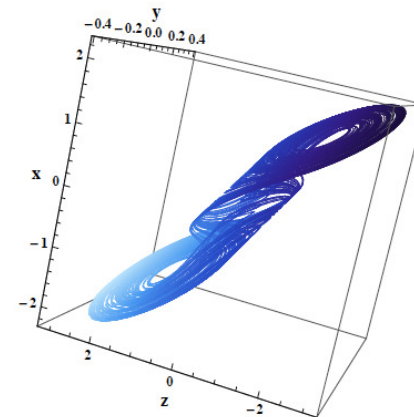
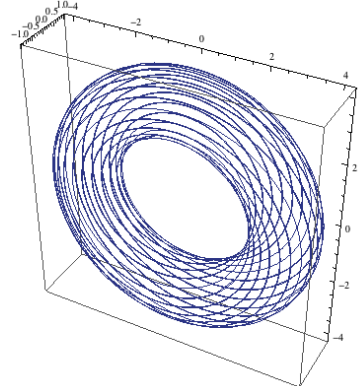
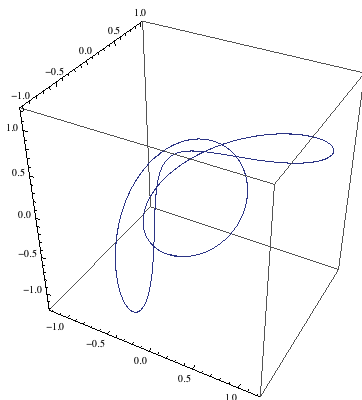
What do I work on?

Nonlinear Dynamical Systems and Embedded Systems

- Applications and Mathematical properties of the Muthuswamy-Chua system
 - Potential Applications to Turbulence Modeling
(MSOE; University of Western Australia, Perth, Australia)
 - Chaotic Hierarchy and Flow Manifolds
(MSOE; University of Western Australia, Perth, Australia; I.U.T. de Toulon, La Garde Cedex, France)
- Applications of Chaotic Delay Differential Equations using Field Programmable Gate Arrays (FPGAs)
(MSOE; Vellore Institute of Technology; University Putra Malaysia, Malaysia; Springer-Verlag)
- Pattern Recognition Using Cellular Neural Networks on FPGAs
(MSOE; Altera Corporation)
- Practical Memristors: discharge tubes, PN junctions and Josephson Junctions
(MSOE; IIT Chennai; University of Western Australia, Perth, Australia;
University of California, Berkeley; Vellore Institute of Technology, Vellore, India)

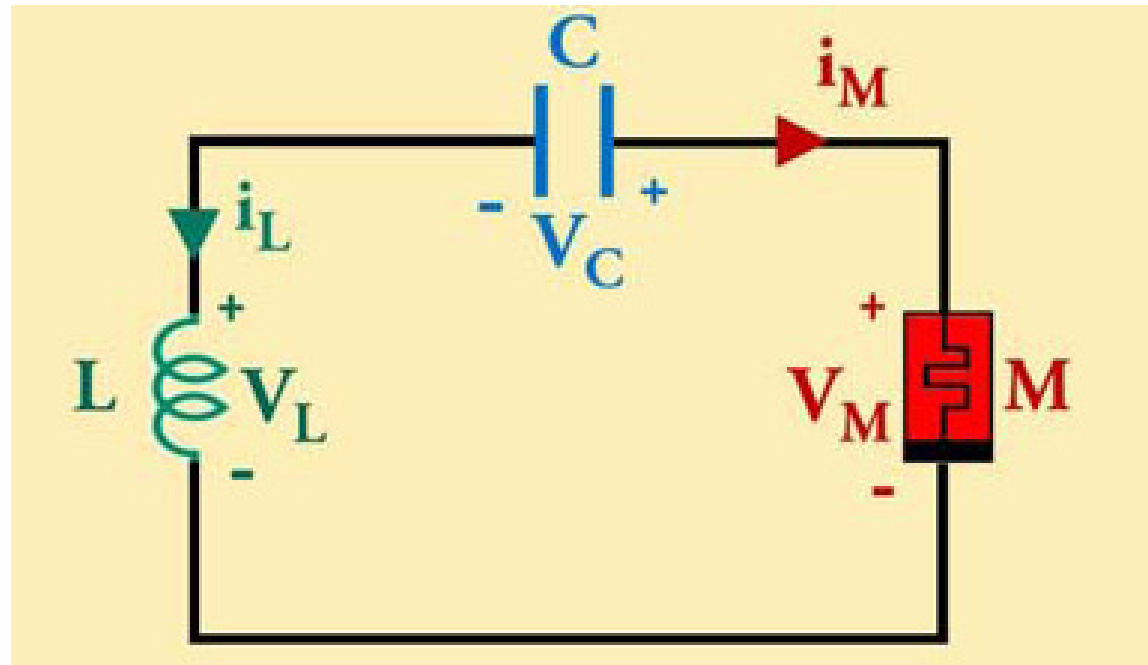
Education

- Nonlinear Dynamics at the undergraduate level (with folks from all over the world ☺)



Goal of This Talk

Obtain chaos in the circuit [5] below:



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- III. The Memristor
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Introduction to chaos

- “Birth” of Chaos: Lorenz Attractor [6]
 - Edward Lorenz introduced the following nonlinear system of differential equations as a crude model of weather in 1963:

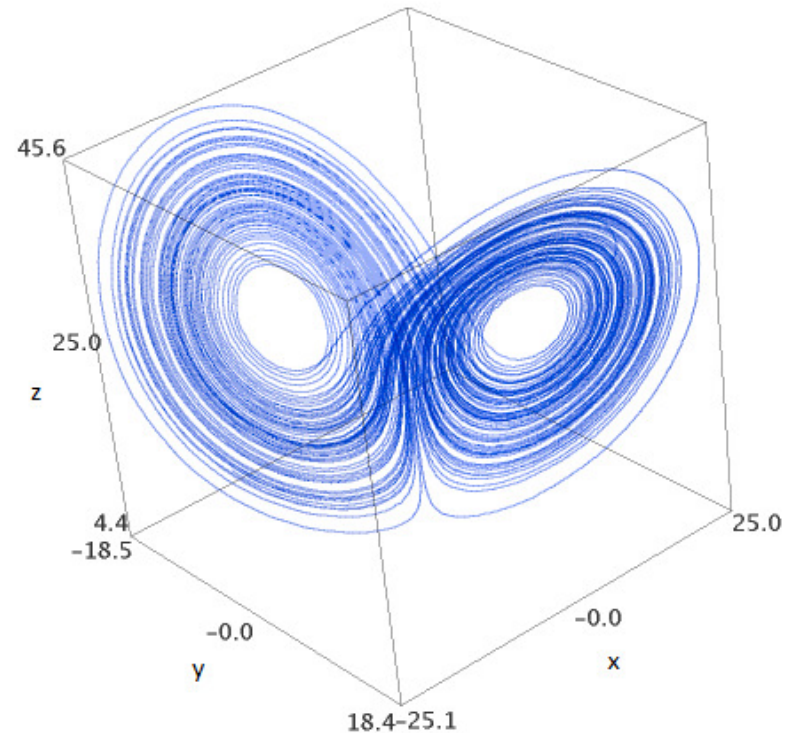
$$\begin{aligned}\dot{\mathbf{x}} &= -\sigma \cdot \mathbf{x} + \sigma \cdot \mathbf{y} \\ \dot{\mathbf{y}} &= \rho \cdot \mathbf{x} - \mathbf{y} - \mathbf{x} \cdot \mathbf{z} \\ \dot{\mathbf{z}} &= -\beta \cdot \mathbf{z} + \mathbf{x} \cdot \mathbf{y}\end{aligned}\quad (1)$$

Parameters: $\sigma = 10, \rho = 28, \beta = \frac{8}{3}$

ICs : $x_0 = 10, y_0 = 20, z_0 = 30,$

Simulation time: 100 seconds

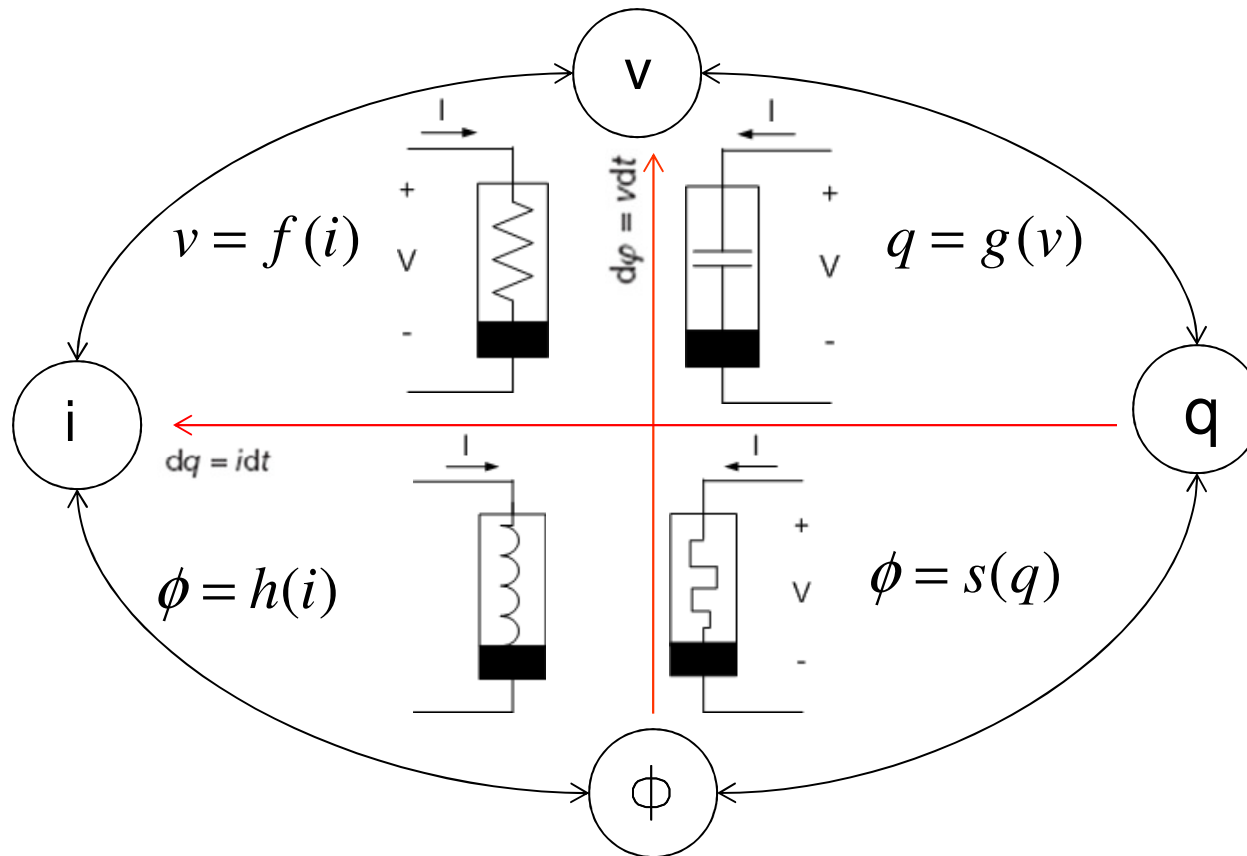
- Lorenz discovered that model dynamics were extremely sensitive to initial conditions and the trajectories were aperiodic but bounded.
- But, does chaos exist *physically*? Answer is: YES. For example, chaotic circuits by Sprott [7].



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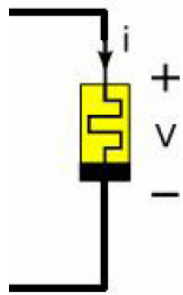
The Fundamental Circuit Elements



Memristors were first postulated by Leon. O Chua in 1971 [2]

Properties of the Memristor [2]

Circuit symbol: A memristor defines a *relation* of the form: $g(\phi, q) = 0$ (2)

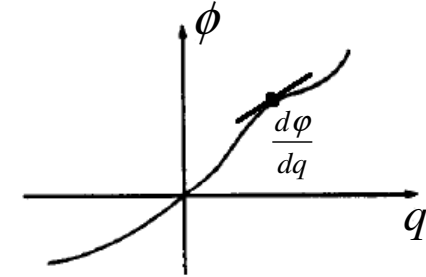


If g is a single-valued function of charge (flux), then the memristor is charge-controlled (flux-controlled)

Memristor i-v relationship:

$M(q(t))$ is the incremental memristance

$$v(t) \triangleq \frac{d\phi}{dt} = \frac{d\phi}{dq} \frac{dq}{dt} \triangleq M(q(t))i(t) \quad (3)$$



Q1: Why is the memristor called “memory resistor”?



Because of the definition of memristance: $v(t) = M(q(t))i(t) = M\left(\int_{-\infty}^t i(\tau)\right)i(t)$

Q2: Why is the memristor not relevant in linear circuit theory?



1. If $M(q(t))$ is a constant: $v(t) = M(q(t))i(t) = Mi(t) = Ri(t)$

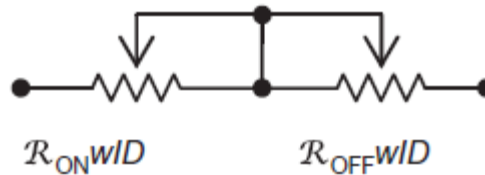
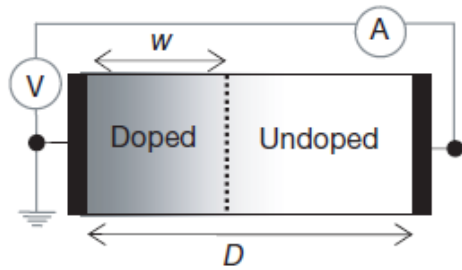
2. Principle of superposition is not* applicable:

$$M\left(\int_{-\infty}^t (i_1 + i_2)(\tau)\right)(i_1 + i_2)(t) = M\left(\int_{-\infty}^t (i_1)(\tau) + \int_{-\infty}^t (i_2)(\tau)\right)(i_1 + i_2)(t) \neq M\left(\int_{-\infty}^t (i_1)(\tau)\right)i_1(t) + M\left(\int_{-\infty}^t (i_2)(\tau)\right)i_2(t)$$

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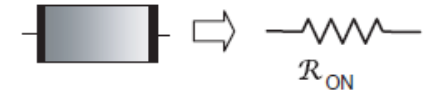
Hewlett-Packard's memristor [9]



Undoped:



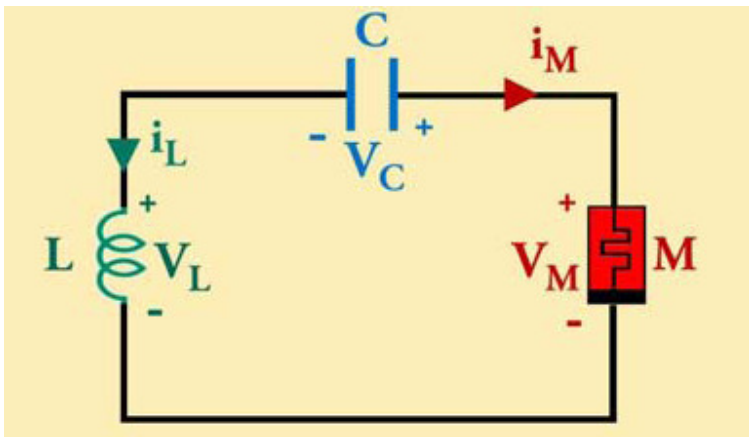
Doped:



$$v(t) = \left(R_{ON} \frac{w(t)}{D} + R_{OFF} \left(1 - \frac{w(t)}{D} \right) \right) i(t)$$

$$\frac{dw(t)}{dt} = \mu_V \frac{R_{ON}}{D} i(t)$$

$$\xrightarrow{R_{ON} \ll R_{OFF}} M(q) = R_{OFF} \left(1 - \frac{\mu_V R_{ON}}{D^2} q(t) \right)$$



Circuit equations:

$$\dot{v}_C = \frac{i_L}{C}$$

$$\dot{q}_M = -i_L \xrightarrow{x \triangleq v_C, z \triangleq q_M, y \triangleq i_L}$$

System equations:

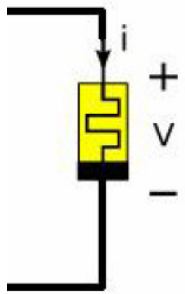
$$\dot{x} = \frac{y}{C}$$

$$\dot{z} = -y \quad (5)$$

$$\dot{y} = \frac{-1}{L} (x + M(z)y)$$

$$v_L + v_C = v_M \Rightarrow L \frac{di_L}{dt} = -v_C + M(q_M) i_M \Rightarrow i'_L = \frac{-1}{L} (v_C + M(q_M) i_L)$$

Memristive Devices [3]



$$\begin{aligned} v &\triangleq R(z, i)i \\ \dot{z} &= f(z, i) \end{aligned} \quad (6)$$

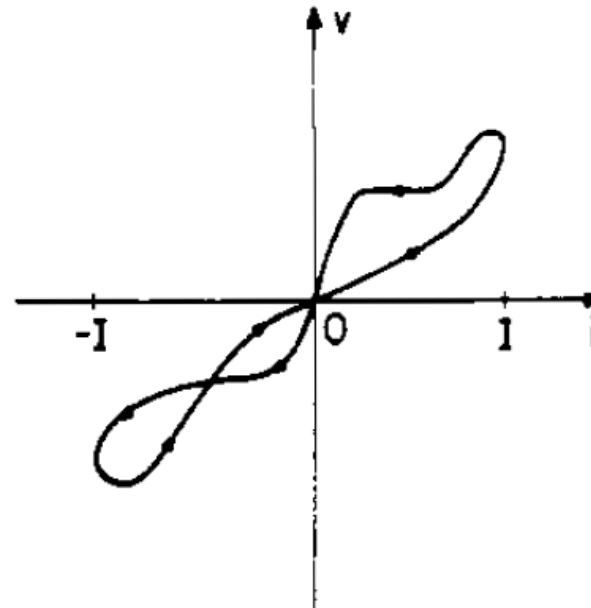
$$\xrightarrow{z \triangleq q, R(z, i) \triangleq M(q)}$$

$$\begin{aligned} v &\triangleq M(q)i \\ \dot{q} &= i \end{aligned}$$

The functions R and f are defined as:

$$R : \mathbb{R}^1 \times \mathbb{R} \rightarrow \mathbb{R}$$

$$f : \mathbb{R}^1 \times \mathbb{R} \rightarrow \mathbb{R}^1$$



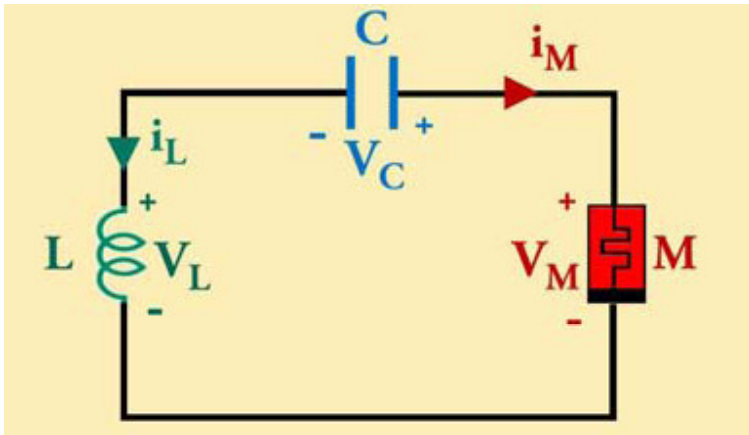
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Derivation of Circuit Equations [5]

$$v_M \triangleq R(z, i_M) i_M$$

$$\dot{z} = f(z, i_M)$$



Circuit equations:

$$\dot{v}_c = \frac{i_L}{C}$$

$$i'_L = \frac{-1}{L}(v_C + R(z, i_L)i_L)$$

$$\dot{z} \triangleq f(z, i_L)$$

System equations:

$$x \triangleq v_c, y \triangleq i_L \quad \xrightarrow{\quad} \quad \dot{x} = \frac{y}{C}$$

$$\dot{y} = \frac{-1}{L}(x + R(z, y)y) \quad (7)$$

$$\dot{z} = f(z, y)$$

Specifically:

$$\dot{x} = \frac{y}{C}$$

$$\dot{y} = \frac{-1}{L}(x + \beta(z^2 - 1)y) \quad (8)$$

$$\dot{z} = -y - \alpha z + yz$$

Parameters:

$$\xrightarrow{\quad} \quad C=1, L=3$$

$$\beta = \frac{3}{2}, \alpha = \frac{3}{5}$$

$$\dot{x} = y$$

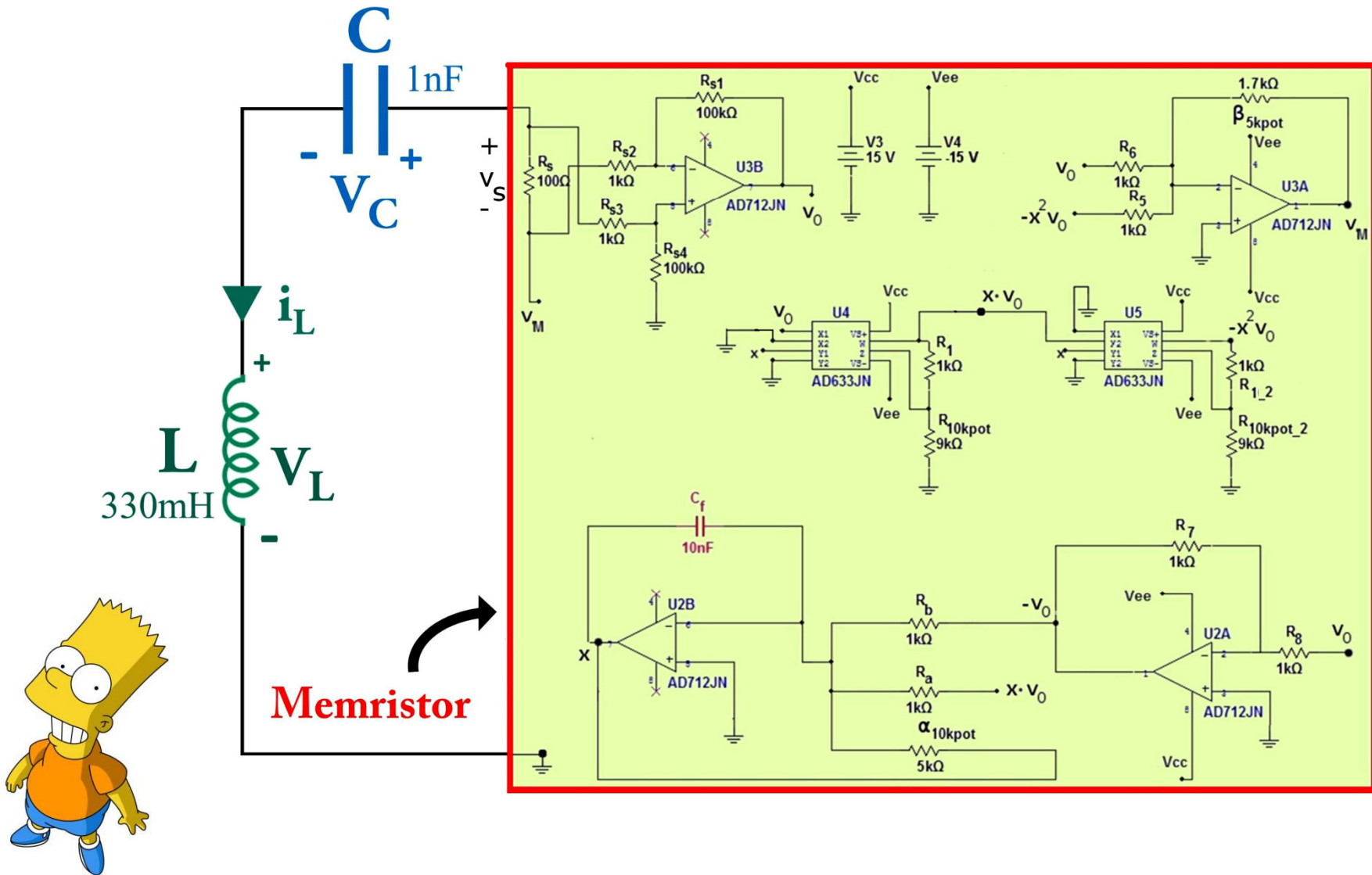
$$\dot{y} = \frac{-x}{3} - \frac{z^2 y}{2} + \frac{y}{2} \quad (9)$$

$$\dot{z} = -y - 0.6z + yz$$

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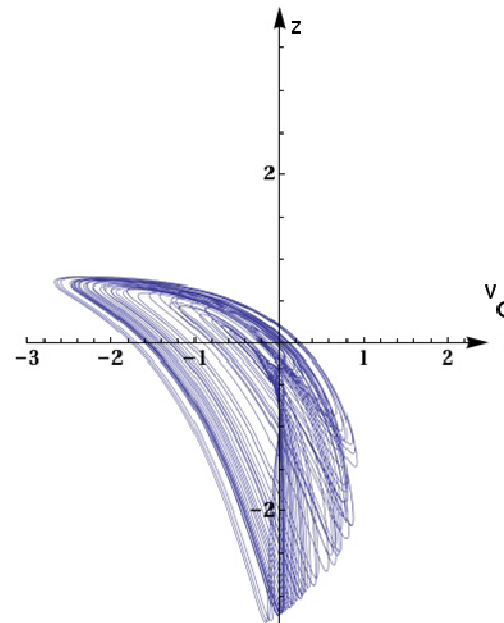
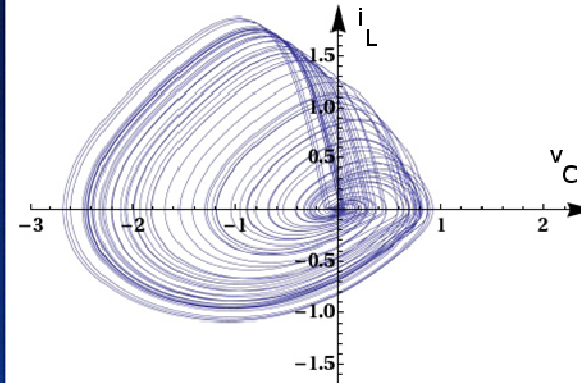
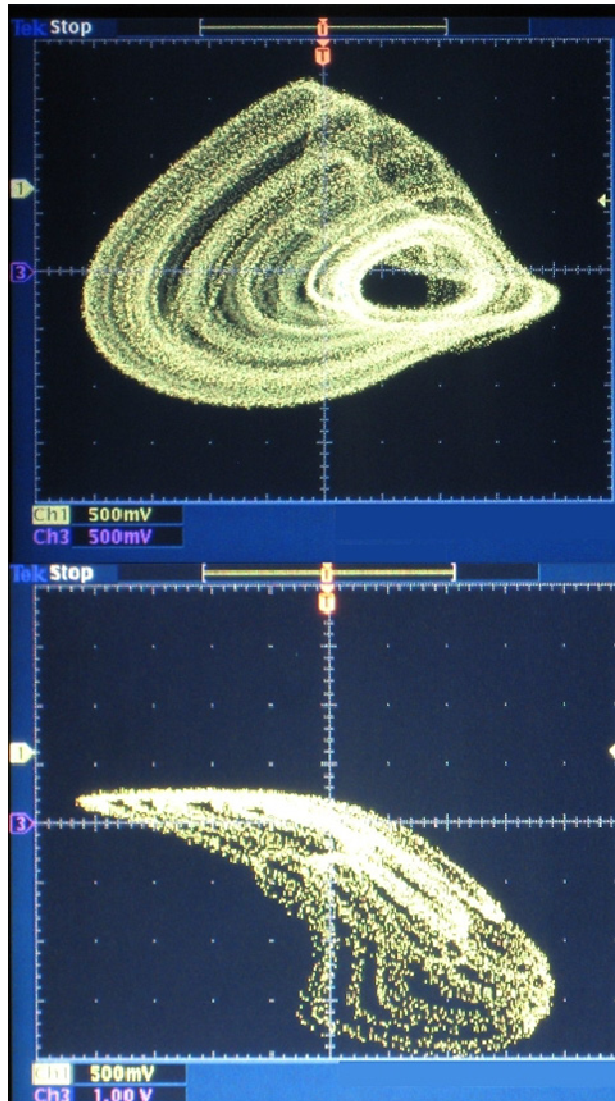
Physical Realization of the Memristor [5]



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Attractors from the Circuit [5]



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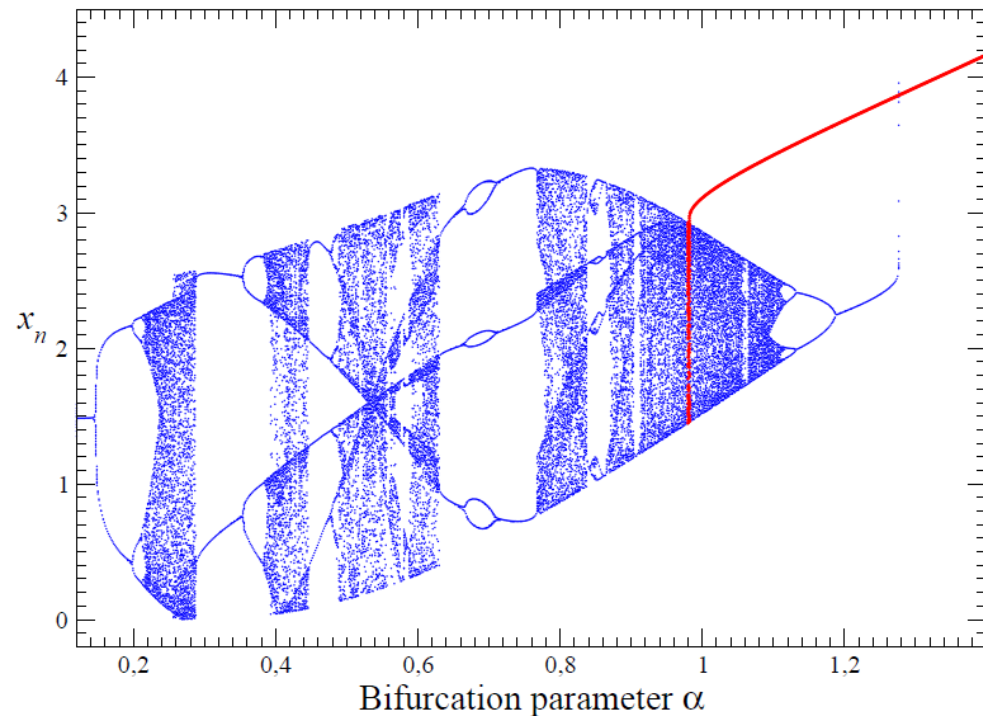
Rigorous Mathematical Analysis

Paper by Ginoux et. al. "Topological Analysis of Chaotic Solution of Three-Element Memristive Circuit". International Journal of Bifurcation and Chaos, Vol. 20, No. 11., pp. 3819 – 3829, Nov. 2010.

$$\dot{x} = y$$

$$\dot{y} = -\frac{x}{3} + \frac{y}{2} - \frac{yz^2}{2}$$

$$\dot{z} = y - \alpha z - yz$$



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Conclusions and Current (future) Work

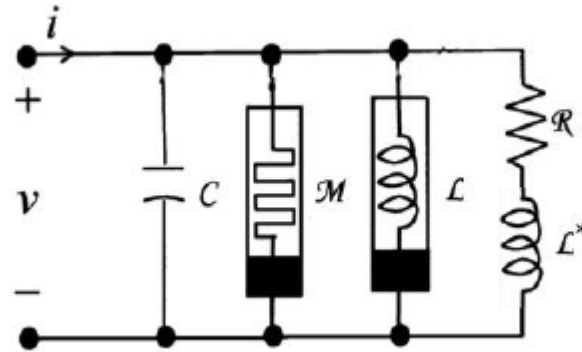
I. Conclusions:

1. We obtained a circuit that uses only *three fundamental circuit elements (only one active)* to obtain chaos.
2. We can pick our choice of nonlinearity, we discussed one particular choice.

II. Current (future) work:

1. **Work with colleagues at the University of Western Australia (related to the ideal memristor: $\cos(\phi)$ term in the Josephson Junction)**

Current (future) Work



The normalized circuit equations are Eqs.(1) through (3)

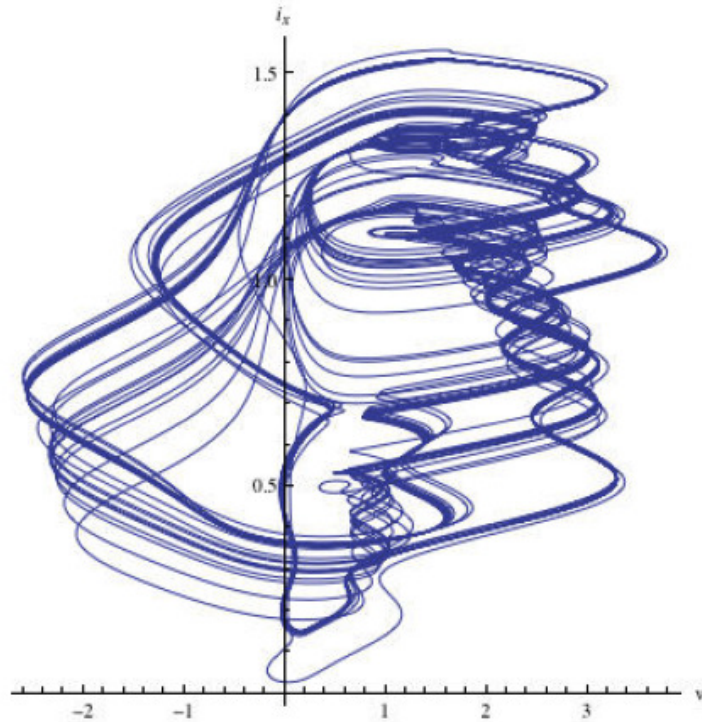
$$\dot{\phi} = v \quad (1)$$

$$\dot{v} = \frac{1}{C} (i - G \cos(k_0 \phi) v - I_0 \sin(\phi) - i_x) \quad (2)$$

$$\dot{i}_x = \frac{1}{L^*} (v - i_x R) \quad (3)$$

The model above was obtained from considering the microscopic theory of Josephson junctions [4]. In Josephson's original papers dealing with thin-film junctions, the coefficients G , k_0 and I_0 in Eq.(2) are dependent on junction voltage [4]. However, this dependence may be neglected provided the voltage stays small when compared to the energy-gap voltage of the individual superconductors [1] Josephson mentions that the \cos term in Eq.(2) contributes to damping effects. Although a similar circuit model was proposed in [2], the shunt inductive branch is not included.

Current (future) Work contd.



References

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Questions?